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## FDC6324L Integrated Load Switch

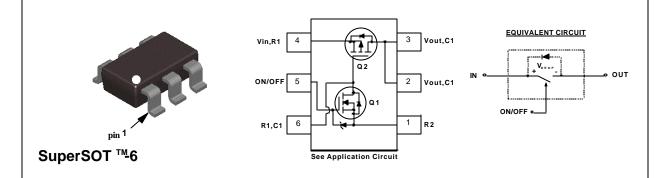
#### **General Description**

These Integrated Load Switches are produced using Fairchild's proprietary, high cell density, DMOS technology. This very high density process is especially tailored to minimize on-state resistance and provide superior switching performance. These devices are particularly suited for low voltage high side load switch application where low conduction loss and ease of driving are needed.

#### **Features**

- High density cell design for extremely low on-resistance.
- V<sub>ONOFF</sub> Zener protection for ESD ruggedness. >6KV Human Body Model.
- SuperSOT<sup>TM</sup>-6 package design using copper lead frame for superior thermal and electrical capabilities.





Absolute Operating Range T<sub>A</sub> = 25°C unless otherwise noted

Symbol	Parameter		FDC6324L	Units
V <sub>IN</sub>	Input Voltage Range		3 - 20	V
V <sub>ON/OFF</sub>	ON/OFF Voltage Range		1.5 - 8	V
I <sub>L</sub>	Load Current @ V <sub>DROP</sub> =0.5V - Continuous	(Note 1)	1.5	A
	- Pulsed	(Note 1 & 3)	2.5	
P <sub>D</sub>	Maximum Power Dissipation	(Note 2a)	0.7	W
$T_J, T_{STG}$	Operating and Storage Temperature Range		-55 to 150	°C
ESD	Electrostatic Discharge Rating MIL-STD-883D Human Body Model (100pf/1500Ohm)		6	kV
THERMA	L CHARACTERISTICS	•		-
$R_{\theta JA}$	Thermal Resistance, Junction-to-Ambient (N	Note 2a)	180	°C/W
$R_{\theta JC}$	Thermal Resistance, Junction-to-Case (N	Note 2)	60	°C/W

Electrical Characteristics (T <sub>A</sub> = 25°C unless otherwise noted)								
Symbol	Parameter	Conditions	Min	Тур	Max	Units		
OFF CHAP	RACTERISTICS							
I <sub>FL</sub>	Forward Leakage Current	V <sub>IN</sub> = 20 V, V <sub>ON/OFF</sub> = 0 V			1	μΑ		
I <sub>RL</sub>	Reverse Leakage Current	V <sub>IN</sub> = -20 V, V <sub>ON/OFF</sub> = 0 V			-1	μΑ		
ON CHAR	ACTERISTICS (Note 3)	·						
V <sub>IN</sub>	Input Voltage		3		20	V		
V <sub>ON/OFF</sub>	On/Off Voltage		1.5		8	V		
V <sub>DROP</sub>	Conduction Voltage Drop @ 1A	V <sub>IN</sub> = 10 V, V <sub>ON/OFF</sub> = 3.3V		0.135	0.2	V		
		V <sub>IN</sub> = 5 V, V <sub>ON/OFF</sub> = 3.3 V		0.215	0.3			
I <sub>L</sub>	Load Current	V <sub>DROP</sub> = 0.2 V, V <sub>IN</sub> = 10 V, V <sub>ON/OFF</sub> = 3.3 V	1			Α		
		V <sub>DROP</sub> = 0.3 V, V <sub>IN</sub> = 5 V, V <sub>ON/OFF</sub> = 3.3 V	1					

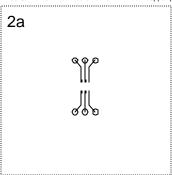
1.  $V_{IN}$ =20V,  $V_{ONOFF}$ =8V,  $V_{DROP}$ =0.5V,  $T_A$ =25°C

2.  $R_{\theta,k}$  is the sum of the junction-to-case and case-to-ambient thermal resistance where the case thermal reference is defined as the solder mounting surface of the drain pins.  $R_{\theta,k}$  is guaranteed by design while  $\mathbf{R}_{\text{\tiny BCA}}$  is determined by the user's board design.

$$P_D(t) = \frac{T_T T_A}{R} = \frac{T_{I-} T_A}{R} = I_D^2(t) \times R_{DSON/@T}$$

 $P_D(t) = \frac{T_T T_A}{R_{\theta,J} A^{\dagger}} = \frac{T_T - T_A}{R_{\theta,J} + R_{\theta,C}(t)} = I_D^2(t) \times R_{D \setminus ON) \otimes T_J}$ Typical  $R_{\theta,JA}$  for single device operation using the board layouts shown below on FR-4 PCB in astill air environment

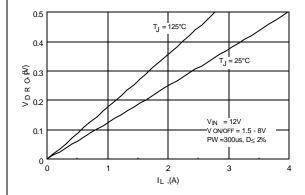
a. 180°C/W when mounted on a 2oz minimum copper pad.



Scale 1 : 1 on letter size paper

3. Pulse Test: Pulse Width ≤ 300µs, Duty Cycle≤ 2.0%

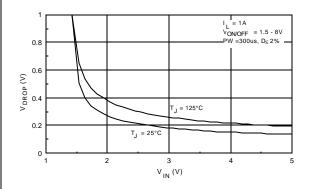
## Typical Electrical Characteristics (T $_{\!\scriptscriptstyle A} = 25~^{\rm o}C$ unless otherwise noted )



0.5
0.4
0.4
0.5
0.6
0.7  $J = 125^{\circ}C$   $T_{J} = 25^{\circ}C$   $T_{J} = 2$ 

Figure 1.  $V_{DROP}$  Versus  $I_L$  at  $V_{IN}$ =12V.





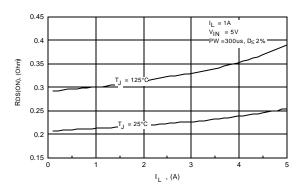


Figure 3.  $V_{DROP}$  Versus  $V_{IN}$  at  $I_L$ =1A.

Figure 4.  $R_{(ON)}$  Versus  $I_L$  at  $V_{IN}$ =5.0V.

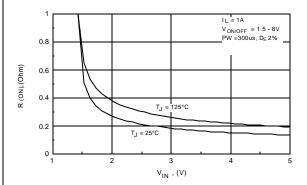


Figure 5. On Resistance Variation with Input Voltage.

## Typical Electrical Characteristics ( $T_A = 25$ $^{\circ}C$ unless otherwise noted )

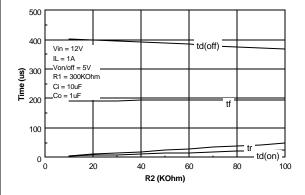


Figure 6. Switching Variation with R2 at Vin=12V and R1=300KOhm.

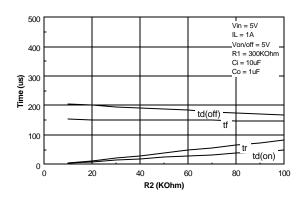


Figure 7. Switching Variation with R2 at Vin=5V and R1=300KOhm.

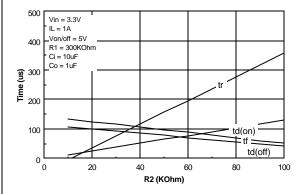


Figure 8. Switching Variation with R2 at Vin=3.3V and R1=300KOhm.

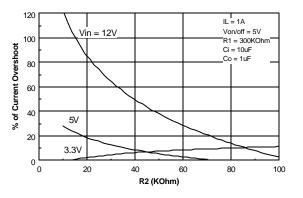


Figure 9. % of Current Overshoot Variation with Vin and R2.

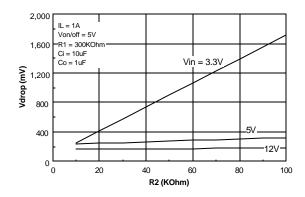


Figure 10. Vdrop Variation with Vin and R2.

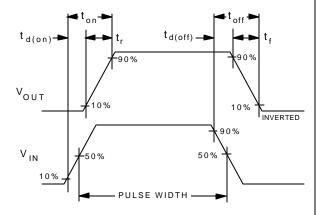


Figure 11. Switching Waveforms.

## Typical Electrical Characteristics (T $_{\!\scriptscriptstyle A} = ~25~^{\rm o}\text{C}$ unless otherwise noted )

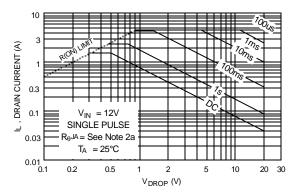


Figure 12. Safe Operating Area.

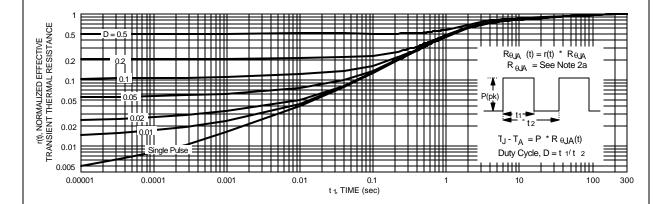
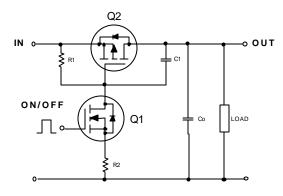


Figure 13. Transient Thermal Response Curve.

Note: Thermal characterization performed on the conditions described in Note 2a. Transient thermal response will change depends on the circuit board

## FDC6324L Load Switch Application

#### **APPLICATION CIRCUIT**



#### **General Description**

This device is particularly suited for computer peripheral switching applications where 20V input and 1A output current capability are needed. This load switch integrates a small N-Channel Power MOSFET (Q1) which drives a large P-Channel Power MOSFET (Q2) in one tiny SuperSOT<sup>TM</sup>-6 package.

A load switch is usually configured for high side switching so that the load can be isolated from the active power source. A P-Channel Power MOSFET, because it does not require its drive voltage above the input voltage, is usually more cost effective than using an N-Channel device in this particular application. A large P-Channel Power MOSFET minimizes voltage drop. By using a small N-Channel device the driving stage is simplified.

### **Component Values**

R1 Typical  $10k - 1M\Omega$ 

 $\begin{array}{ccc} R2 & Typical & 0 - 10k\Omega & \text{(optional)} \\ C1 & Typical & 1000pF & \text{(optional)} \end{array}$ 

#### **Design Notes**

- R1 is needed to turn off Q2.
- R2 can be used to soft start the switch in the case the output capacitance Co is small.
- $R2 \le$  should be at least 10 times smaller than R1 to guarantee Q1 turns on.
- By using R1 and R2 a certain amount of current is lost from the input. This bias current loss is given by the equation

 $I_{BIAS\_LOSS} = \frac{Vin}{R1 + R2}$  when the switch is ON.  $I_{BIAS\_LOSS}$  can be minimized by large R1.

• R2 and C<sub>RSS</sub> of Q2 make ramp for slow turn on. If excessive overshoot current occurs due to fast turn on, additional capacitance C1 can be added externally to slow down the turn on.

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FACT™ QFET™ FACT Quiet Series™ QS™

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