Precision, Micropower, Single Supply Instrumentation Amplifier (Fixed Gain = 10 or 100)

FEATURES

- Gain Error: 0.04% Max
- Gain Nonlinearity: 0.0008% (8ppm) Max
- Gain Drift: 4ppm/°C Max Supply Current: 105μA Max
- Offset Voltage: 160µV Max
- Offset Voltage Drift: 0.4μV/°C Typ
- Offset Current: 600pA Max CMRR, G = 100: 100dB Min
- 0.1Hz to 10Hz Noise: 0.9µVp-p Typ
 - 2.3pAp-p Typ
- Gain Bandwidth Product: 250kHz Min
- Single or Dual Supply Operation
- Surface Mount Package Available

APPLICATIONS

- Differential Signal Amplification in Presence of Common Mode Voltage
- Micropower Bridge Transducer Amplifier
 - Thermocouples
 - Strain Gauges
 - Thermistors
- Differential Voltage-to-Current Converter
- Transformer Coupled Amplifier
- 4mA to 20mA Bridge Transmitter

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DESCRIPTION

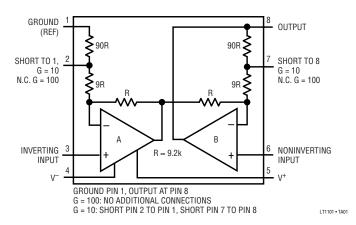
The LT®1101 establishes the following milestones: (1) It is the first micropower instrumentation amplifier, (2) It is the first single supply instrumentation amplifier, (3) It is the first instrumentation amplifier to feature fixed gains of 10 and/or 100 in low cost, space-saving 8-lead packages.

The LT1101 is completely self-contained: no external gain setting resistor is required. The LT1101 combines its micropower operation (75µA supply current) with a gain error of 0.008%, gain linearity of 3ppm, gain drift of 1ppm/°C. The output is guaranteed to drive a 2k load to ±10V with excellent gain accuracy.

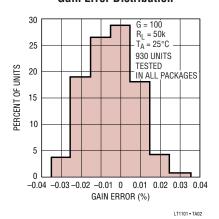
Other precision specifications are also outstanding: 50 µV input offset voltage, 130 pA input offset current, and low drift $(0.4\mu\text{V/}^{\circ}\text{C} \text{ and } 0.7\text{pA/}^{\circ}\text{C})$. In addition, unlike other instrumentation amplifiers, there is no output offset voltage contribution to total error.

A full set of specifications are provided with ±15V dual supplies and for single 5V supply operation. The LT1101 can be operated from a single lithium cell or two Ni-Cad batteries. Battery voltage can drop as low as 1.8V, yet the LT1101 still maintains its gain accuracy. In single supply applications, both input and output voltages swing to within a few millivolts of ground. The output sinks current while swinging to ground—no external, power consuming pull down resistors are needed.

TYPICAL APPLICATION



Gain Error Distribution



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ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage	±22V
Differential Input Voltage	
Input Voltage Equal to Positive Su	ipply Voltage
10V Below Negative Su	pply Voltage
Output Short Circuit Duration	Indefinite

Operating Temperature Range

LT1101AM/LT1101M (OBSOLETE) ... -55°C to 125°C

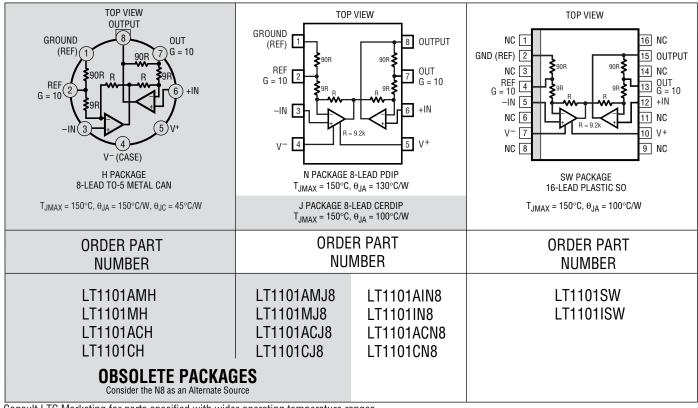
LT1101AI/LT1101I -40°C to 85°C

LT1101AC/LT1101C 0°C to 70°C

Storage Temperature Range -65°C to 150°C

Lead Temperature (Soldering, 10 sec) 300°C

PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS $V_S = 5V$, 0V, $V_{CM} = 0.1V$, $V_{REF(PIN 1)} = 0.1V$, G = 10 or 100, $T_A = 25^{\circ}C$, unless otherwise noted. (Note 4)

			LT1	LT1101AM/AI/AC			LT1101M/I/C		
SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
G _E	Gain Error	$G = 100, V_0 = 0.1V \text{ to } 3.5V, R_L = 50k$ $G = 10, V_0 = 0.1V \text{ to } 3.5V, R_L = 50k$		0.010 0009	0.050 0.040		0.011 0.010	0.075 0.060	% %
G _{NL}	Gain Nonlinearity	G = 100, R _L = 50k G = 10, R _L = 50k (Note 2)		20 3	60 7		20 3	75 8	ppm ppm
V _{OS}	Input Offset Voltage	LT1101SW		50	160		60 250	220 600	μV μV
I _{OS}	Input Offset Current			0.13	0.60		0.15	0.90	nA
I_{B}	Input Bias Current			6	8		6	10	nA
I _S	Supply Current			75	105		78	120	μА
					-		-		1101fa



ELECTRICAL CHARACTERISTICS $V_S = 5V$, 0V, $V_{CM} = 0.1V$, $V_{REF(PIN~1)} = 0.1V$, G = 10 or 100, $T_A = 25^{\circ}C$, unless otherwise noted. (Note 4)

			LT1	101AM/A	I/AC	LT1101M/I/C			
SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
CMRR	Common Mode Rejection Ratio	1k Source Imbalance G = 100, V _{CM} = 0.07V to 3.4V G = 10, V _{CM} = 0.07V to 3.1V	95 84	106 100		92 82	105 99		dB dB
	Minimum Supply Voltage	(Note 5)		1.8	2.3		1.8	2.3	V
$\overline{V_0}$	Maximum Output Voltage Swing	Output High, 50k to GND Output High, 2k to GND Output Low, V _{REF} = 0, No Load Output Low, V _{REF} = 0, 2k to GND Output Low, V _{REF} = 0, I _{SINK} = 100µA	4.1 3.5	4.3 3.9 3.3 0.5 90	6 1 130	4.1 3.5	4.3 3.9 3.3 0.5 90	6 1 130	V V mV mV
BW	Bandwidth	G = 100 (Note 2) G = 10 (Note 2)	2.0 22	3.0 33		2.0 22	3.0 33		kHz kHz
SR	Slew Rate	(Note 2)	0.04	0.07		0.04	0.07		V/µs

$V_S=\pm 15 V,\, V_{CM}=0 V,\, T_A=25^{\circ}C,\, Gain=10$ or 100, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	LT ⁻	1101AM/A Typ	I/AC Max	MIN	.T1101M/I/ Typ	C Max	UNITS
GE	Gain Error	G = 100, $V_0 = \pm 10V$, $R_L = 50k$ G = 100, $V_0 = \pm 10V$, $R_L = 2k$ G = 100, $V_0 = \pm 10V$, $R_L = 50k$ or $2k$	"""	0.008 0.011 0.008	0.040 0.055 0.040		0.009 0.012 0.009	0.060 0.070 0.060	% % %
G _{NL}	Gain Nonlinearity	G = 100, R _L = 50k G = 100, R _L = 2k G = 10, R _L = 50k or 2k		7 24 3	16 45 8		8 25 3	20 60 9	ppm ppm ppm
V _{OS}	Input Offset Voltage	LT1101SW		50	160		60 250	220 600	μV μV
I _{OS}	Input Offset Current			0.13	0.60		0.15	0.90	nA
I_{B}	Input Bias Current			6	8		6	10	nA
	Input Resistance Common Mode Differential Mode	(Note 2) (Note 2)	4 7	7 12		3 5	7 12		GΩ GΩ
e _n	Input Noise Voltage	0.1Hz to 10Hz (Note 3)		0.9	1.8		0.9		μVр-р
	Input Noise Voltage Density	f ₀ = 10Hz (Note 3) f ₀ = 1000Hz (Note 3)		45 43	64 54		45 43		nV/√ <u>Hz</u> nV/√Hz
e _n	Input Noise Current	0.1Hz to 10Hz (Note 3)		2.3	4.0		2.3		рАр-р
	Input Noise Current Density	f ₀ =10Hz (Note 3) f ₀ = 1000Hz		0.06 0.02	0.10		0.06 0.02		pA/√ <u>Hz</u> pA/√Hz
	Input Voltage Range	G = 100 G = 10	13.0 -14.4 11.5 -13.0	13.8 -14.7 12.5 -13.3		13.0 -14.4 11.5 -13.0	13.8 -14.7 12.5 -13.3		V V V
CMRR	Common Mode Rejection Ratio	1k Source Imbalance G = 100, Over CM Range G = 10, Over CM Range	100 84	112 100		98 82	112 99		dB dB
PSRR	Power Supply Rejection Ratio	$V_S = +2.2V, -0.1V \text{ to } \pm 18V$	102	114		100	114		dB
Is	Supply Current			92	130		94	150	μА
	1	-	1			1			1101fa



ELECTRICAL CHARACTERISTICS $V_S = \pm 15 V$, $V_{CM} = 0 V$, $T_A = 25 ^{\circ} C$, Gain = 10 or 100, unless otherwise noted.

			LT1101AM/AI			LT1101M/I				
SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNITS	
V_0	Maximum Output Voltage Swing	R _L = 50k R _L = 2k	13.0 11.0	14.2 13.2		13.0 11.0	14.2 13.2		V	
BW	Bandwidth	G = 100 (Note 2) G = 10 (Note 2)	2.3 25	3.5 37		2.3 25	3.5 37		kHz kHz	
SR	Slew Rate		0.06	0.10		0.06	0.10		V/µs	

ELECTRICAL CHARACTERISTICS $V_S = \pm 15 V$, $V_{CM} = 0 V$, Gain = 10 or 100, $-55^{\circ}C \le T_A \le 125^{\circ}C$ for AM/M grades, $-40^{\circ}C \le T_A \le 85^{\circ}C$ for Al/I grades, unless otherwise noted.

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SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
G _E	Gain Error	$G = 100, V_0 = \pm 10V, R_L = 50k$ $G = 100, V_0 = \pm 10V, R_L = 5k$ $G = 10, V_0 = \pm 10V, R_L = 50k \text{ or } 5k$		0.024 0.030 0.015	0.070 0.100 0.070		0.026 0.035 0.018	0.100 0.130 0.100	% % %
TCG _E	Gain Error Drift (Note 2)	G = 100, R _L = 50k G = 100, R _L = 5k G = 10, R _L = 50k or 5k		2 2 1	4 7 4		2 2 1	5 8 5	ppm/°C ppm/°C ppm/°C
G _{NL}	Gain Nonlinearity	G = 100, R _L = 50k G = 100, R _L = 5k G = 10, R _L = 50k G = 10, R _L = 5k		24 70 4 10	70 300 13 40		26 75 5 12	90 500 15 60	ppm ppm ppm ppm
V _{OS}	Input Offset Voltage	LT1101ISW		90	350		110 110	500 950	μV μV
$\Delta V_{0S}/\Delta T$	Input Offset Voltage Drift	(Note 2) LT1101ISW		0.4	2.0		0.5 0.5	2.8 4.8	μV/°C mV/°C
I _{OS}	Input Offset Current			0.16	0.80		0.19	1.30	nA
$\Delta I_{OS}/\Delta T$	Input Offset Current Drift	(Note 2)		0.5	4.0		0.8	7.0	pA/°C
I _B	Input Bias Current			7	10		7	12	nA
$\Delta I_B/\Delta T$	Input Bias Current Drift	(Note 2)		10	25		10	30	pA/°C
CMRR	Common Mode Rejection Ratio	G = 100, V _{CM} = -14.4V to 13V G = 100, V _{CM} = -13V to 11.5V	96 80	111 99		94 78	111 98		dB dB
PSRR	Power Supply Rejection Ratio	$V_S = 3.0, -0.1V \text{ to } \pm 18V$	98	110		94	110		dB
Is	Supply Current			105	165		108	190	μΑ
$\overline{V_0}$	Maximum Output Voltage Swing	R _L = 50k R _L = 5k	12.5 11.0	14.0 13.5		12.5 11.0	14.0 13.5		V

ELECTRICAL CHARACTERISTICS $V_S=\pm 15V,\ V_{CM}=0V,\ Gain=10\ or\ 100,\ 0^{\circ}C \le T_A \le 70^{\circ}C,\ unless$ otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	LT1101AC TYP	MAX	MIN	LT1101C/S TYP	MAX	UNITS
G _E	Gain Error	$G = 100, V_0 = \pm 10V, R_L = 50k$ $G = 100, V_0 = \pm 10V, R_L = 2k$ $G = 10, V_0 = \pm 10V, R_L = 50k$ or 2k		0.012 0.018 0.009	0.055 0.085 0.055		0.014 0.020 0.010	0.080 0.100 0.080	% % %
TCGE	Gain Error Drift (Note 2)	G = 100, R _L = 50k G = 100, R _L = 2k G = 10, R _L = 50k or 5k		1 2 1	4 7 4		1 2 1	5 9 5	ppm/°C ppm/°C ppm/°C
G _{NL}	Gain Nonlinearity	$G = 100, R_L = 50k$ $G = 100, R_L = 2k$ $G = 10, R_L = 50k$ or $2k$		9 33 4	25 75 10		10 36 4	35 100 11	ppm ppm ppm
V _{OS}	Input Offset Voltage	LT1101SW		70	250		85 300	350 800	μV μV
$\Delta V_{0S}/\Delta T$	Input Offset Voltage Drift	(Note 2) LT1101SW		0.4	2.0		0.5 1.2	2.8 4.5	μV/°C μV/°C
I _{OS}	Input Offset Current			0.14	0.70		0.17	1.10	nA
$\Delta I_{0S}/\Delta T$	Input Offset Current Drift	(Note 2)		0.5	4.0		0.8	7.0	pA/°C
I_B	Input Bias Current			6	9		6	11	nA
$\Delta I_B/\Delta T$	Input Bias Current Drift	(Note 2)		10	25		10	30	pA/°C
CMRR	Common Mode Rejection Ratio	G = 100, V _{CM} = -14.4V to 13V G = 100, V _{CM} = -13V to 11.5V	98 82	112 100		96 80	112 99		dB dB
PSRR	Power Supply Rejection Ratio	$V_S = 2.5, -0.1V \text{ to } \pm 18V$	100	112		97	112		dB
Is	Supply Current			98	148		100	170	μΑ
$\overline{V_0}$	Maximum Output Voltage Swing	R _L = 50k R _L = 2k	±12.5 ±10.5	±14.1 ±13.0		±12.5 ±10.5	±14.1 ±13.0		V

ELECTRICAL CHARACTERISTICS $V_S = 5V$, 0V, $V_{CM} = 0.1V$, $V_{REF(PIN 1)} = 0.1V$, Gain = 10 or 100, $-40^{\circ}C \le T_A \le 85^{\circ}C$ for Al/I grades, unless otherwise noted (Note 4).

			LT1101AM/AI			LT1101M/I			
SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
G _E	Gain Error	G = 100, V ₀ = 0.1V to 3.5V, R _L = 50k G = 10, V _{CM} = 0.15, R _L = 50k		0.026 0.011	0.080 0.070		0.028 0.014	0.120 0.100	% %
TCGE	Gain Error Drift	R _L = 50k (Note 2)		1	4		1	5	ppm/°C
G _{NL}	Gain Nonlinearity	G = 100, R _L = 50k G = 10, R _L = 50k (Note 2)		45 4	110 13		48 5	140 15	ppm ppm
V _{OS}	Input Offset Voltage	LT1101ISW		90	350		110 110	500 950	μV μV
$\Delta V_{0S}/\Delta T$	Input Offset Voltage Drift	(Note 2) LT1101ISW		0.4	2.0		0.5 0.5	2.8 4.8	μV/°C μV/°C
I _{OS}	Input Offset Current			0.16	0.80		0.19	1.30	nA
$\Delta V_{OS}/\Delta T$	Input Offset Current Drift	(Note 2)		0.5	4.0		0.8	7.0	pA/°C
I _B	Input Bias Current			7	10		7	12	nA
$\Delta I_B/\Delta T$	Input Bias Current Drift	(Note 2)		10	25		10	30	pA/°C
CMRR	Common Mode Rejection Ratio	G = 100, V _{CM} = 0.1V to 3.2V G = 10, V _{CM} = 0.1V to 2.9V, V _{REF} = 0.15V	91 80	105 98		88 77	104 97		dB dB
Is	Supply Current			88	135		92	160	μА
V_0	Maximum Output Voltage Swing	Output High, 50k to GND Output High, 2k to GND Output Low, V _{REF} = 0, No Load Output Low, V _{REF} = 0, 2k to GND Output Low, V _{REF} = 0, I _{SINK} = 100µA	3.8 3.0	4.1 3.7 4.5 0.7 125	8 1.5 170	3.8 3.0	4.1 3.7 4.5 0.7 125	8 1.5 170	V V mV mV

ELECTRICAL CHARACTERISTICS $V_S = 5V$, 0V, $V_{CM} = 0.1V$, $V_{REF(PIN~1)} = 0.1V$, Gain = 10 or 100, $0^{\circ}C \le T_A \le 70^{\circ}C$, unless otherwise noted (Note 4).

SYMBOL	PARAMETER	CONDITIONS	MIN	LT1101AC Typ	MAX	MIN	LT1101C/S TYP	MAX	UNITS
G _E	Gain Error	G = 100, V ₀ = 0.1V to 3.5V, R _L = 50k G = 10, V _{CM} = 0.15V, R _L = 50k		0.017 0.010	0.065 0.060		0.018 0.012	0.095 0.080	% %
TCGE	Gain Error Drift	R _L = 50k (Note 2)		1	4		1	5	ppm/°C
G _{NL}	Gain Nonlinearity	G = 100, R _L = 50k G = 10, R _L = 50k (Note 2)		25 4	80 10		25 4	100 11	ppm ppm
V _{OS}	Input Offset Voltage	LT1101SW		70	250		85 300	350 800	μV μV
$\Delta V_{0S}/\Delta T$	Input Offset Voltage Drift	(Note 2) LT1101SW		0.4	2.0		0.5 1.2	2.8 4.5	μV/°C μV/°C
I _{OS}	Input Offset Current			0.14	0.70		0.17	1.10	nA
$\Delta I_{OS}/\Delta T$	Input Offset Current Drift	(Note 2)		0.5	4.0		0.8	7	pA/°C
I_{B}	Input Bias Current			6	9		6	11	nA
$\Delta I_B/\Delta T$	Input Bias Current Drift	(Note 2)		10	25		10	30	pA/°C
CMRR	Common Mode Rejection Ratio	G = 100, V _{CM} = 0.07V to 3.3V G = 10, V _{CM} = 0.07V to 3V, V _{REF} = 0.15V	93 82	105 99		90 80	104 98		dB dB
Is	Supply Current			80	120		85	145	μА
V_0	Maximum Output Voltage Swing	Output High, 50k to GND Output High, 2k to GND Output Low, V _{REF} = 0, No Load Output Low, V _{REF} = 0, 2k to GND Output Low, V _{REF} = 0, I _{SINK} = 100μA	4.0 3.3	4.2 3.8 4 0.6 100	7 1.2 150	4.0 3.3	4.2 3.8 4 0.6 100	7 1.2 150	V V mV mV

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

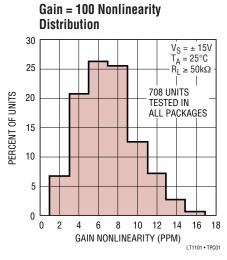
Note 2: This parameter is not tested. It is guaranteed by design and by inference from other tests.

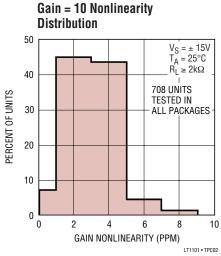
Note 3: This parameter is tested on a sample basis only.

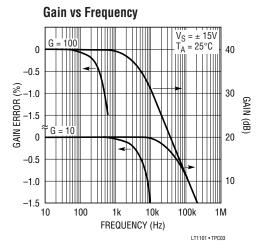
Note 4: These test conditions are equivalent to $V_S = 4.9V$, -0.1V, $V_{CM} = 0V$, $V_{REF(PIN1)} = 0V$.

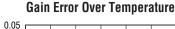
Note 5: Minimum supply voltage is guaranteed by the power supply rejection test. The LT1101 actually works at 1.8V supply with minimal degradation in performance.

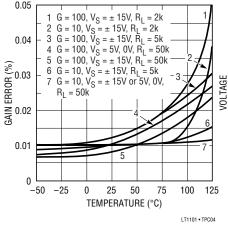




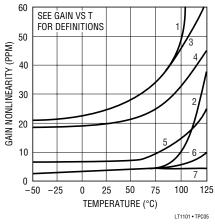




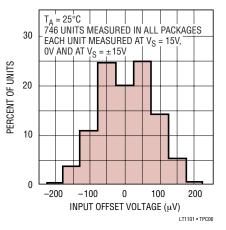




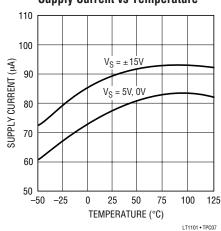




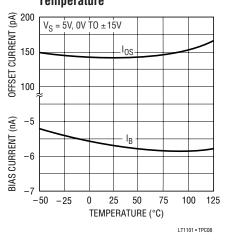
Input Offset Voltage Distribution



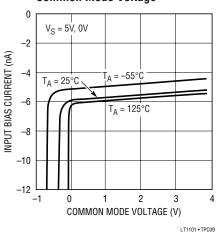
Supply Current vs Temperature



Input Bias and Offset Currents vs Temperature



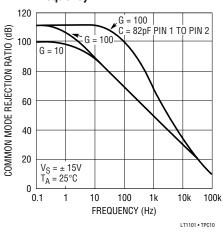
Input Bias Current vs Common Mode Voltage



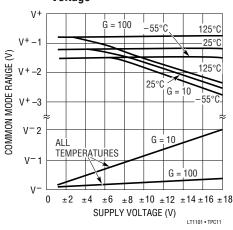
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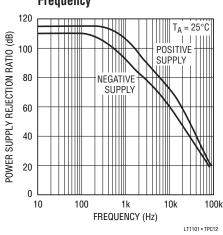
Common Mode Rejection Ratio vs Frequency



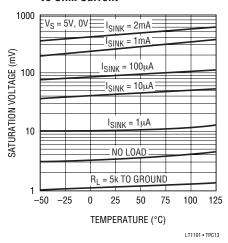
Common Mode Range vs Supply Voltage



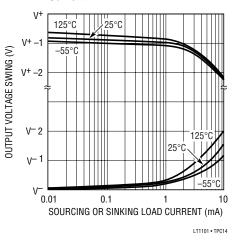
Power Supply Rejection Ratio vs Frequency



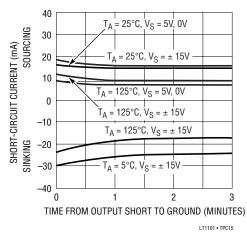
Output Saturation vs Temperature vs Sink Current



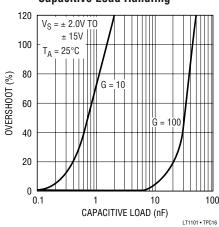
Output Voltage Swing vs Load Current



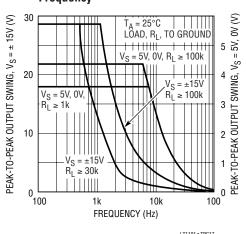
Short-Circuit Current vs Time



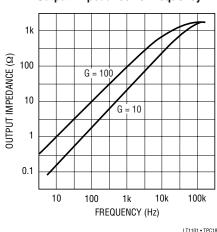
Capacitive Load Handling



Undistorted Output Swing vs Frequency

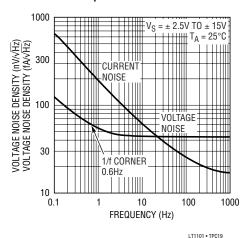


Output Impedance vs Frequency

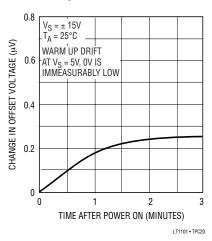


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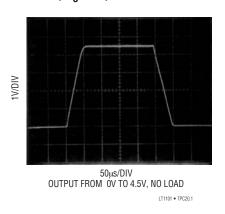
Noise Spectrum



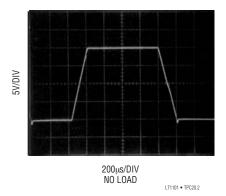
Warm-Up Drift



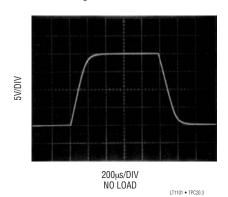
Large Signal Transient Response G = 10, $V_S = 5V$, 0V



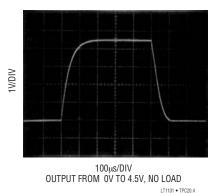
Large Signal Transient Response G = 10, $V_S = 15V$



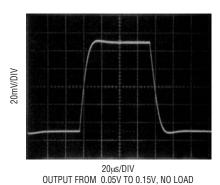
Large Signal Transient Response G = 100, $V_S = \pm 15V$



Large Signal Transient Response $G = 100, V_S = 5V, 0V$

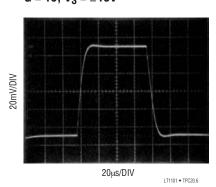


Small Signal Transient Response G = 10, $V_S = 5V$, 0V

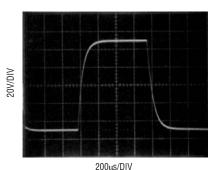


T11101 • TPC20.5

Small Signal Transient Response G = 10, $V_S = \pm 15V$



Small Signal Transient Response $G = 100, V_S = 5V, 0V$

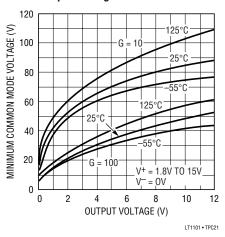


OUTPUT FROM 0.05V TO 0.15V, NO LOAD (RESPONSE WITH $V_S = \pm 15V$, G = 100 IS IDENTICAL)

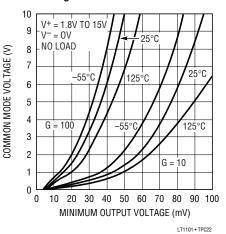
LT1101 • TPC20.7



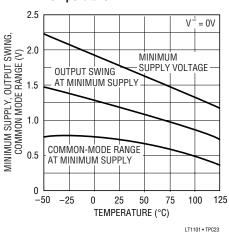
Single Supply: Minimum Common Mode Voltage vs Output Voltage



Single Supply: Minimum Output Voltage vs Common Mode Voltage



Minimum Supply Voltage vs Temperature



APPLICATIONS INFORMATION

Single Supply Applications

The LT1101 is the first instrumentation amplifier which is fully specified for single supply operation, (i.e. when the negative supply is 0V). Both the input common mode range and the output swing are within a few millivolts of ground.

Probably the most common application for instrumentation amplifiers is amplifying a differential signal from a transducer or sensor resistance bridge. All competitive instrumentation amplifiers have a minimum required common mode voltage which is 3V to 5V above the negative supply. This means that the voltage across the bridge has to be 6V to 10V or dual supplies have to be used (i.e., micropower) single battery usage is not attainable on competitive devices.

The minimum output voltage obtainable on the LT1101 is a function of the input common mode voltage. When the common mode voltage is high and the output is low, current will flow from the output of amplifier A into the output of amplifier B. See the Minimum Output Voltage vs Common Mode Voltage plot.

Similarly, the Single Supply Minimum Common Mode Voltage vs Output Voltage plot specifies the expected common mode range.

When the output is high and input common mode is low, the output of amplifier A has to sink current coming from the output of amplifier B. Since amplifier A is effectively in unity gain, its input is limited by its output.

Common Mode Rejection vs Frequency

The common mode rejection ratio (CMRR) of the LT1101 starts to roll off at a relatively low frequency. However, as shown on the Common Mode Rejection Ratio vs Frequency plot, CMRR can be enhanced significantly by connecting an 82pF capacitor between pins 1 and 2. This improvement is only available in the gain 100 configuration, and it is in excess of 30dB at 60Hz.

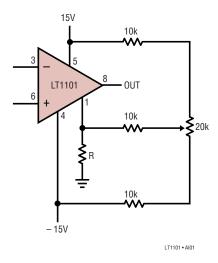
Offset Nulling

The LT1101 is not equipped with dedicated offset null terminals. In many bridge transducer or sensor applications, calibrating the bridge simultaneously eliminates the instrumentation amplifier's offset as a source of error. For example, in the Micropower Remote Temperature Sensor Application shown, one adjustment removes the offset errors due to the temperature sensor, voltage reference and the LT1101.

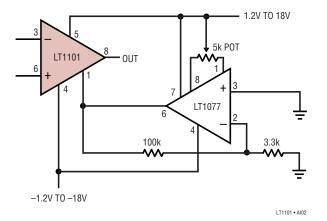


APPLICATIONS INFORMATION

A simple resistive offset adjust procedure is shown below. If $R = 5\Omega$ for G = 10, and $R = 50\Omega$ for G = 100, then the effect of R on gain error is approximately 0.006%. Unfortunately, about $450\mu A$ has to flow through R to bias the reference terminal (Pin 1) and to null out the worst-case offset voltage. The total current through the resistor network can exceed 1mA, and the micropower advantage of the LT1101 is lost.



Another offset adjust scheme uses the LT1077 micropower op amp to drive the reference Pin 1. Gain error and common mode rejection are unaffected, the total current increase is $45\mu A$. The offset of the LT1077 is trimmed and amplified to match and cancel the offset voltage of the LT1101. Output offset null range is $\pm 25 \text{mV}$.



Gains Between 10 and 100

Gains between 10 and 100 can be achieved by connecting two equal resistors (= R_X) between Pins 1 and 2 and Pins 7 and 8.

$$Gain = 10 + \frac{R_{\chi}}{R + R_{\chi}/90}$$

The nominal value of R is $9.2k\Omega$. The usefulness of this method is limited by the fact that R is not controlled to better than $\pm 10\%$ absolute accuracy in production. However, on any specific unit, 90R can be measured between Pins 1 and 2.

Input Protection

Instrumentation amplifiers are often used in harsh environments where overload conditions can occur. The LT1101 employs PNP input transistors, consequently the differential input voltage can be $\pm 30V$ (with $\pm 15V$ supplies, $\pm 36V$ with $\pm 18V$ supplies) without an increase in input bias current. Competitive instrumentation amplifiers have NPN inputs which are protected by back-to-back diodes. When the differential input voltage exceeds $\pm 1.3V$ on these competitive devices, input current increases to the milliampere level; more than $\pm 10V$ differential voltage can cause permanent damage.

When the LT1101's inputs are pulled above the positive supply, the inputs will clamp a diode voltage above the positive supply. No damage will occur if the input current is limited to 20mA.

 500Ω resistors in series with the inputs protect the LT1101 when the inputs are pulled as much as 10V below the negative supply.

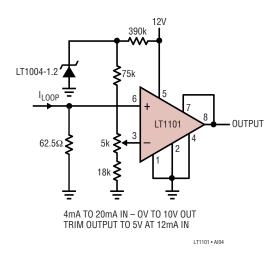
TECHNOLOGY TECHNOLOGY

APPLICATIONS INFORMATION

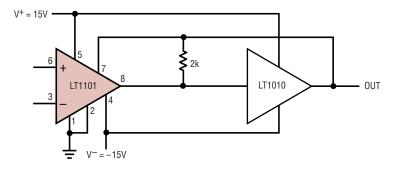
Micropower, Battery Operated Remote Temperature Sensor

3V REMOTE TEMP SENSOR LM134-3 T1004-1.2 **₹**75k \$2210Ω \$62Ω 2k POT LT1101 OUT 6 G = 1010mV/°C 10<u>0nA</u> **≤**10k **≨** 20k °K TRIM OUTPUT TO 250mV AT 25°C TEMPERATURE RANGE = 2.5°C TO 150°C ACCURACY = ± 0.5 °C LT1101 • AI03

4mA to 20mA Loop Receiver

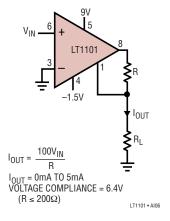


Instrumentation Amplifier with ± 150 mA Output Current



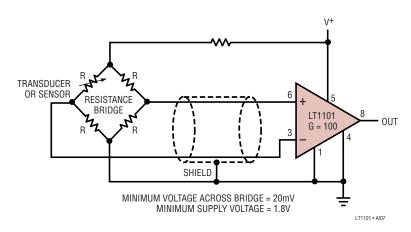
GAIN = 10, DEGRADED BY 0.01% DUE TO LT1010 OUTPUT = $\pm 10V$ INTO 75Ω (TO 1.5kHz) DRIVES ANY CAPACITIVE LOAD SINGLE SUPPLY APPLICATION (V+ = 5V, V⁻ = 0V): $V_{OUT\ MIN}$ = 120mV, $V_{OUT\ MAX}$ = 3.4V

Voltage Controlled Current Source

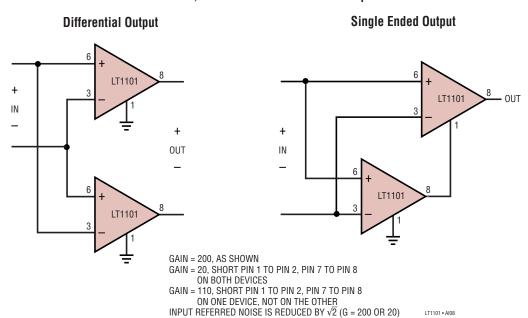


APPLICATIONS INFORMATION

Differential Voltage Amplification from a Resistance Bridge



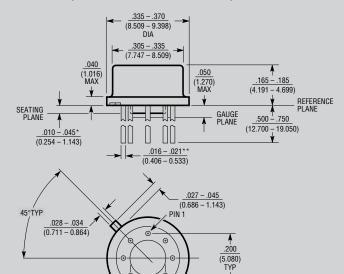
Gain = 20, 110 or 200 Instrumentation Amplifier



PACKAGE DESCRIPTION

H Package 8-Lead TO-5 Metal Can (.200 Inch PCD)

(Reference LTC DWG # 05-08-1320)

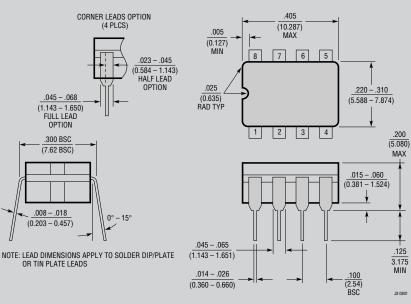


*LEAD DIAMETER IS UNCONTROLLED BETWEEN THE REFERENCE PLANE AND THE SEATING PLANE

.110 - .160 (2.794 - 4.064) INSULATING STANDOFF

J8 Package 8-Lead CERDIP (Narrow .300 Inch, Hermetic)

(Reference LTC DWG # 05-08-1110)



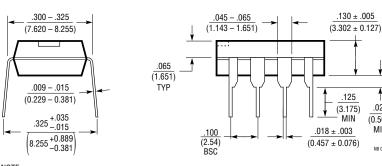
OBSOLETE PACKAGES

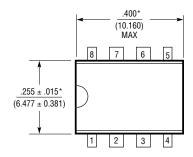
^{**}FOR SOLDER DIP LEAD FINISH, LEAD DIAMETER IS $\frac{.016 - .024}{(0.406 - 0.610)}$

PACKAGE DESCRIPTION

N8 Package 8-Lead PDIP (Narrow .300 Inch)

(Reference LTC DWG # 05-08-1510)





INCHES

NOTE: INCHES

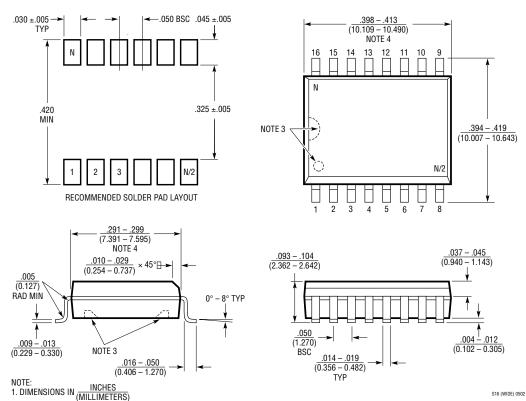
1. DIMENSIONS ARE $\frac{\text{INCHES}}{\text{MILLIMETERS}}$ *THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .010 INCH (0.254mm)

SW Package 16-Lead Plastic Small Outline (Wide .300 Inch)

.020

(0.508)

(Reference LTC DWG # 05-08-1620)



2. DRAWING NOT TO SCALE

2. DIAWAND TO SOAGE.

3. PIN 1 IDENT, NOTCH ON TOP AND CAVITIES ON THE BOTTOM OF PACKAGES ARE THE MANUFACTURING OPTIONS.

THE PART MAY BE SUPPLIED WITH OR WITHOUT ANY OF THE OPTIONS.

4. THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.

MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .006" (0.15mm)

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